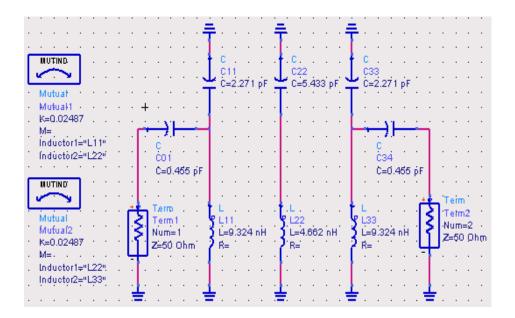
# **Magnetically Coupled Chebychev BPF**

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This sheet is used to design a **microwave magnetically coupled BPF**. The maths for this filter is derived from equations in "Theory & Design of Microwave Filters", Ian Hunter, IEE Press.

#### Main user input area:

$$\mathbf{\mu} := 10^{-6} \cdot \mathbf{m} \qquad \mathbf{n} \mathbf{H} := 10^{-9} \cdot \mathbf{henry}$$

$L_{ar_{db}} \coloneqq 10$	Passband return loss in dB Order of the filter, this needs to be above 2			
<u>N:= 3</u>				
$f_{gm} \coloneqq 1 \cdot GHz$	bandwidth := $20 \cdot MHz$	$Z_0 := 50 \cdot 0$		

### Start Calculating...

$$f_{low} := f_{gm} - \frac{bandwidth}{2} \qquad f_{high} := f_{gm} + \frac{bandwidth}{2} \qquad bw := \frac{f_{high} - f_{low}}{f_{gm}}$$
$$f_{bp}(f) := \frac{1}{bw} \cdot \left(\frac{f}{f_{gm}} - \frac{f_{gm}}{f}\right) \cdot f_{gm} \qquad bw percent := bw \cdot 100 \qquad \mathbf{a} := \frac{f_{gm}}{bandwidth} \qquad bw percent = 2$$
$$\mathbf{a} = 50$$

### LPF Frequency Response, for Chebychev Polynomials

This section plots the frequency response for the Chebychev BPF

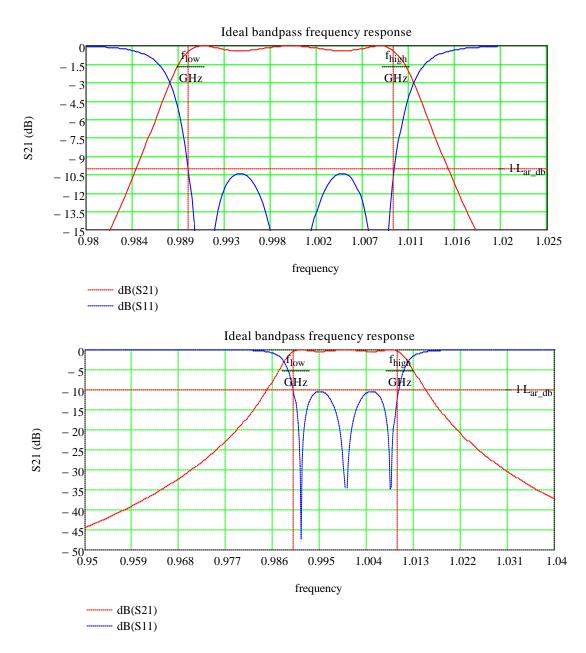
$$Y_{o} \coloneqq \frac{1}{Z_{o}} \qquad \qquad \underbrace{\mathbf{e}}_{\mathsf{M}} \coloneqq \left(\frac{2 \cdot L_{ar\_db}}{10} - 1\right)^{-0.5}$$

**e** = 0.1005

$$\begin{split} L_{A}(f,f_{c}) &\coloneqq \left| 10 \cdot \log \Biggl[ 1 + \mathbf{e} \cdot \left( \cos \Biggl( N \cdot \arcsin \Biggl( \frac{f}{f_{c}} \Biggr) \Biggr) \Biggr)^{2} \Biggr] & \text{if } f \leq f_{c} \\ 10 \cdot \log \Biggl[ \Biggl[ 1 + \mathbf{e} \cdot \Biggl( \cosh \Biggl( N \cdot \arcsin \Biggl( \frac{f}{f_{c}} \Biggr) \Biggr) \Biggr)^{2} \Biggr] \Biggr] & \text{if } f > f_{c} \\ S11_{A}(f,f_{1}) &\coloneqq 10 \cdot \log \Biggl[ 1 - 10 \Biggl( \frac{-L_{A}(f,f_{1})}{10} \Biggr) \Biggr] \end{split}$$

 $f\_bp\_narrow \coloneqq 0.99 \cdot f_{low}, \frac{f_{high} - f_{low}}{100} + 0.99 \cdot f_{low}.. \ 1.01 \cdot f_{high}$ 

 $f\_bp\_wide \coloneqq (f_{gm} - 3 \cdot f_{gm} \cdot bw), (f_{gm} - 3 \cdot f_{gm} \cdot bw) + \frac{f_{gm} \cdot bw}{100} ... f_{gm} + (3 \cdot f_{gm} \cdot bw)$ 

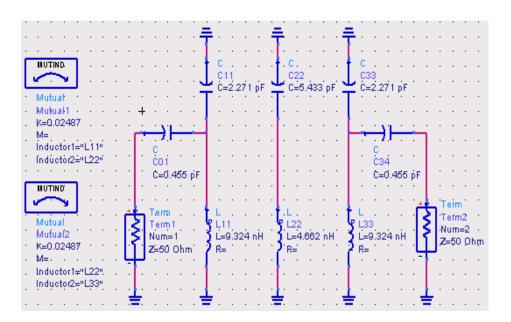


## Calculate the circuit values.....

$$\begin{aligned} \mathbf{P} &:= \sinh\left(\frac{1}{N} \cdot \operatorname{asinh}\left(\frac{1}{\mathbf{e}}\right)\right) & n := 1, 2.. N \\ nn &:= 1, 2.. N - 1 \\ \mathbf{K}_{nn, nn+1} := \frac{\left[\mathbf{P}^{2} + \left(\sin\left(\frac{nn \cdot \mathbf{p}}{N}\right)\right)^{2}\right]^{0.5}}{\mathbf{P}^{2}} \\ \mathbf{Q}_{n} &:= \frac{2}{\mathbf{P}} \sin\left[\frac{(2 \cdot n - 1) \cdot \mathbf{p}}{2 \cdot N}\right] \\ \operatorname{Cap}_{0,1} &:= \frac{1}{\left[2 \cdot \mathbf{p} \cdot f_{gm} \cdot (\mathbf{a} - 1)^{0.5} \cdot Z_{o}\right]} \\ \operatorname{Cap}_{N, N+1} &:= \frac{1}{\left[2 \cdot \mathbf{p} \cdot f_{gm} \cdot (\mathbf{a} - 1)^{0.5} \cdot Z_{o}\right]} \\ \operatorname{Cap}_{N, N+1} &:= \frac{1}{\left[2 \cdot \mathbf{p} \cdot f_{gm} \cdot (\mathbf{a} - 1)^{0.5} \cdot Z_{o}\right]} \\ \operatorname{Cap}_{N, N+1} &:= \frac{1}{\left[2 \cdot \mathbf{p} \cdot f_{gm} \cdot (\mathbf{a} - 1)^{0.5} \cdot Z_{o}\right]} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} \\ \operatorname{Cap}_{N, N+1} &:= \frac{2}{C_{n} \cdot 2 \cdot \mathbf{p} \cdot f_{gm}} \\ \operatorname{Cap}_{N, N+1} \\ \operatorname{Cap}_{N, N$$

$$\operatorname{Cap}_{n,n} \coloneqq \frac{\operatorname{C}_{n}}{2 \cdot \mathbf{p} \cdot f_{gm} \cdot Z_{o}} \qquad \operatorname{Cap}_{1,1} \coloneqq \frac{\operatorname{C}_{1}}{2 \cdot \mathbf{p} \cdot f_{gm} \cdot Z_{o}} - \frac{(\mathbf{a} - 1)^{0.5}}{2 \cdot \mathbf{p} \cdot f_{gm} \cdot Z_{o} \cdot \mathbf{a}} \qquad \operatorname{Cap}_{N,N} \coloneqq \operatorname{Cap}_{1,1}$$

$\operatorname{Cap}_{n,n} =$		$\operatorname{Ind}_{n,n} =$		$\frac{K_{nn, nn+1}}{m} =$
2.271	· pF	9.324	· nH	a
5.433		4.662		0.02487
2.271		9.324		0.02487



#### **Simulation Results**

To test these values I have run a simulation on ADS for N=3, Return Loss 10dB, BW=20MHz, Zo=50R. You can see the results are in good agreement. Both the ripple and return loss (S11) agree.

You can see the downside of these filters that the frequency response repeats and the out of band attenuation is not a good we would like and not as good as expected from the ideal chebychev response.

